

In the Claims

1 1. (currently amended) A method for detecting symbols of a modulated
2 signal received via channels of a wireless communications system,
3 comprising:
4 obtaining an initial estimate of a symbol transmitted via the channels
5 from a previous channel estimate and a received symbol;
6 updating the channel estimate;
7 optimizing the a next estimate of the transmitted symbol which
8 maximizes an expectation of a log likelihood function by averaging a
9 logarithm of a likelihood function over unknown parameters h of the
10 channels;
11 quantizing the next estimate of the transmitted symbol;
12 comparing the quantized next estimate of the transmitted symbol with
13 the ~~previous~~ initial estimate of the transmitted symbol to determine if the
14 ~~previous~~ initial estimate of the transmitted symbol and the quantized next
15 estimate of the transmitted symbol have converged; and otherwise
16 inputting the quantized next estimate of the transmitted symbol as the
17 initial estimate of the transmitted symbol; and
18 repeating the updating, the optimizing, the quantizing, and the
19 comparing until the ~~previous~~ initial estimate of the transmitted symbol and
20 the quantized next estimate of the transmitted symbol converge.

1 2. (currently amended) The method of claim 1 wherein the modulated signal
2 is a MPSK modulated signal having a positive constant equivalent to an
3 energy of the modulated signal, and using only phase information during the
4 updating.

1 3. (currently amended) The method of claim 1 wherein the comparing
2 further comprises:
3 subtracting the ~~previous~~ initial estimate of the transmitted symbol
4 from the quantized next estimate of the transmitted symbol to obtain a
5 difference; and
6 determining that the ~~previous~~ initial estimate and the quantized next
7 estimate have converged when an absolute value of the difference is less
8 than a predetermined threshold.

1 4. (currently amended) The method of claim 1 further comprising:
2 obtaining the initial estimate of the transmitted symbol from the
3 channel estimate of a pilot symbol received via the channels.

1 5. (currently amended) The method of claim 1 further comprising:
2 obtaining the initial estimate of the transmitted symbol from the
3 channel estimate of a previously received symbol.

1 6. (currently amended) The method of claim 1 wherein the optimizing
2 further comprises:
3 using only a fast Fourier transform matrix, the received ~~signal~~ symbol,
4 and the previous channel estimate .

7. (currently amended) The method of claim 1 wherein the next estimate of the transmitted symbol is quantized according to ~~the signal constellation~~ a constellation of the received signal.

8. (currently amended) The method of claim 1 further comprising:
determining a posterior covariance matrix Ξ_p $\hat{\Sigma}_p$ of the channels using a FFT matrix \mathbf{W} , the ~~previous~~ initial estimate of the transmitted symbol \mathbf{X}_p , the received symbol \mathbf{Y} , and a Gaussian noise variance σ^2 as

$$\hat{\Sigma}_p = (\mathbf{W}^H \mathbf{X}_p^H \mathbf{X}_p \mathbf{W} / \sigma^2)^{-1},$$

determining a posterior mean \hat{h}_p of a channel impulse response as

$$\hat{h}_p = \hat{\Sigma}_p (\mathbf{W}^H \mathbf{X}_p^H \mathbf{Y} / \sigma^2);$$

determining a channel update coefficients matrix \mathbf{C} for recovering the next estimate of the transmitted symbol; and

applying the coefficient matrix \mathbf{C} to the posterior mean \hat{h}_p , the FFT matrix \mathbf{W} , and the received ~~signal~~ symbol \mathbf{Y} according to

$$\tilde{\mathbf{X}}_{p+1} = \mathbf{C}^{-1} (\hat{h}_p^H \mathbf{W}^H \mathbf{Y})^T \text{ to optimize the next estimate of the } \underline{\text{transmitted symbol}}$$

$$\mathbf{X}_{p+1} \underline{\tilde{\mathbf{X}}_{p+1}}.$$

9. (currently amended) The method of claim 1 further comprising:
determining a posterior covariance matrix Ξ_p $\hat{\Sigma}_p$ of the channels using a FFT matrix \mathbf{W} , the previous estimate of the transmitted symbol \mathbf{X}_p , a channel convergence matrix Σ^{-1} , and a Gaussian noise variance σ^2 as

$$\hat{\Sigma}_p = (\mathbf{W}^H \mathbf{X}_p^H \mathbf{X}_p \mathbf{W} / \sigma^2 + \Sigma^{-1})^{-1};$$

6 determining a posterior mean \hat{h}_p of a channel impulse response as
7 $\hat{h}_p = \hat{\Sigma}_p (\mathbf{W}^H \mathbf{X}_p^H \mathbf{Y} / \sigma^2 + \Sigma^{-1} E\{h\})$, where the received symbol is \mathbf{Y} , and $E\{h\}$ is
8 a channel impulse response;
9 determining a channel update coefficients matrix \mathbf{C} for recovering the
10 next estimate of the transmitted symbol; and
11 applying the coefficient matrix \mathbf{C} to the posterior mean \hat{h}_p , the FFT
12 matrix \mathbf{W} , and the received ~~signal~~ symbol \mathbf{Y} according to
13 $\tilde{\mathbf{X}}_{p+1} = \mathbf{C}^{-1} (\hat{h}_p^H \mathbf{W}^H \mathbf{Y})^T$ to optimize the next estimate of the transmitted symbol
14 ~~\mathbf{X}_{p+1}~~ $\tilde{\mathbf{X}}_{p+1}$.

1 10. (cancelled)

1 11. (original) The method of claim 1 further comprising:
2 modulating the signal using orthogonal frequency division
3 multiplexing.

1 12. (currently amended) A system for detecting symbols of a modulated
2 signal received via a plurality of channel of a wireless communications
3 system, comprising:
4 means for obtaining an initial estimate of a symbol transmitted via the
5 channels;
6 means for updating the channel estimate;
7 means for optimizing a next estimate of the transmitted symbol which
8 maximizes an expectation of a log likelihood function by averaging a
9 logarithm of a likelihood function over unknown parameters h of the

10 channels;
11 means for quantizing the next estimate of the transmitted symbol;
12 means for comparing the quantized next estimate of the transmitted
13 symbol with the previous estimate of the transmitted symbol to determine if
14 the ~~previous~~ initial estimate and the quantized next estimate have converged;
15 and otherwise
16 means for making the quantized next estimate of the transmitted
17 symbol an input for a next iteration; and
18 means for repeating the updating, the optimizing, the quantizing, and
19 comparing until the ~~previous~~ initial estimate of the transmitted symbol and
20 the quantized next estimate of the transmitted symbol converge.

1 13. (currently amended) The system of claim 12 wherein the modulated
2 signal is a ~~MPSK~~ multiple phase shift keying modulated signal having a
3 positive constant equivalent to an energy of the modulated signal, and using
4 only phase information during the updating.

1 14. (currently amended) The system of claim 12 further comprising:
2 means for subtracting the ~~previous~~ initial estimate of the transmitted
3 symbol from the quantized next estimate of the transmitted symbol to obtain
4 a difference; and
5 means for determining that the ~~previous~~ initial estimate and the next
6 estimate have converged when an absolute value of the difference is less
7 than a predetermined threshold.

15. (currently amended) The system of claim 12 wherein the initial estimate of the transmitted symbol is obtained from a pilot symbol received via the channels.

16. (currently amended) The system of claim 12 wherein the initial estimate of the transmitted symbol is obtained from a channel estimate from ~~the~~ previous a previously received symbol.

17. (currently amended) The system of claim 12 further comprising:
means for determining a posterior covariance matrix Σ_p $\hat{\Sigma}_p$ of the channels using a FFT matrix \mathbf{W} , the initial estimate of the ~~previous~~ transmitted symbol \mathbf{X}_p , the received symbol \mathbf{Y} , and a Gaussian noise variance σ^2 as $\hat{\Sigma}_p = (\mathbf{W}^H \mathbf{X}_p^H \mathbf{X}_p \mathbf{W} / \sigma^2)^{-1}$,
means for determining a posterior mean \hat{h}_p of the channel impulse response as $\hat{h}_p = \hat{\Sigma}_p (\mathbf{W}^H \mathbf{X}_p^H \mathbf{Y} / \sigma^2)$;
means for determining a channel update coefficients matrix \mathbf{C} for recovering the next estimate of the ~~next~~ transmitted symbol; and
means for applying the coefficient matrix \mathbf{C} to the posterior mean \hat{h}_p , the FFT matrix \mathbf{W} , and the received ~~signal~~ symbol \mathbf{Y} according to $\tilde{\mathbf{X}}_{p+1} = \mathbf{C}^{-1} (\hat{h}_p^H \mathbf{W}^H \mathbf{Y})^T$ to maximize the next estimate of the ~~next~~ symbol \mathbf{X}_{p+1} $\tilde{\mathbf{X}}_{p+1}$.

18. (currently amended) The system of claim 12 further comprising:

means for determining a posterior covariance matrix Σ_p of the

channels using the FFT matrix \mathbf{W} , the initial estimate of the ~~previous~~

transmitted symbol \mathbf{X}_p , a ~~channels~~ channel convergence matrix Σ^{-1} , and a

Gaussian noise variance σ^2 as $\hat{\Sigma}_p = (\mathbf{W}^H \mathbf{X}_p^H \mathbf{X}_p \mathbf{W} / \sigma^2 + \Sigma^{-1})^{-1}$;

means for determining a posterior mean \hat{h}_p of ~~the channels~~ a channel

impulse response as $\hat{h}_p = \hat{\Sigma}_p (\mathbf{W}^H \mathbf{X}_p^H \mathbf{Y} / \sigma^2 + \Sigma^{-1} E\{\underline{h}\})$, where the received

symbol is \mathbf{Y} and $E\{\underline{h}\}$ is a channel impulse response;

means for determining a channel update coefficients matrix \mathbf{C} for

recovering the estimate of the next transmitted symbol; and

means for applying the coefficient matrix \mathbf{C} to the posterior mean \hat{h}_p ,

the FFT matrix \mathbf{W} , and the received ~~signal~~ symbol \mathbf{Y} according to

$\tilde{\mathbf{X}}_{p+1} = \mathbf{C}^{-1} (\hat{h}_p^H \mathbf{W}^H \mathbf{Y})^T$ to maximize the next estimate of the ~~next~~ symbol ~~\mathbf{X}_{p+1}~~

$\tilde{\mathbf{X}}_{p+1}$.

19. (cancelled)

20. (original) The system of claim 12 wherein the signal is modulated using

orthogonal frequency division multiplexing.